

FEATURES

- Lowpass Filter with No DC Error
- Low Passband Noise
- Operates DC to 20kHz
- Operates on a Single 5V Supply or Up to $\pm 8V$
- 5th Order Filter
- Maximally Flat Response
- Internal or External Clock
- Cascadable for Faster Roll-off
- Buffer Available
- 8 Pin DIP Package

APPLICATIONS

- 60Hz Lowpass Filters
- Anti-Aliasing Filter
- Low Level Filtering
- Rolling Off AC Signals from High DC Voltages
- Digital Voltmeters
- Scales
- Strain Gauges

DESCRIPTION

The LTC1062 is a 5th order all pole maximally flat lowpass filter with no DC error. Its unusual architecture puts the filter outside the DC path so DC offset and low frequency noise problems are eliminated. This makes the LTC1062 very useful for lowpass filters where DC accuracy is important.

The filter input and output are simultaneously taken across an external resistor. The LTC1062 is coupled to the signal through an external capacitor. This R,C reacts with the internal switched capacitor network to form a 5th order rolloff at the output.

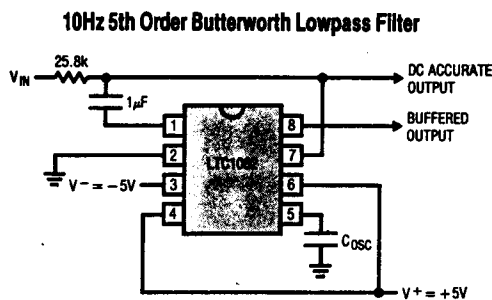
The filter cutoff frequency is set by an internal clock which can be externally driven. The clock to cutoff frequency ratio is typically 100:1, allowing the clock ripple to be easily removed.

Two LTC1062s can be cascaded to form a 10th order quasi max flat lowpass filter. The device can be operated with single or dual supplies ranging from $\pm 2.5V$ to $\pm 9V$.

The LTC1062 is manufactured using Linear Technology's enhanced LTCMOS™ silicon gate process.

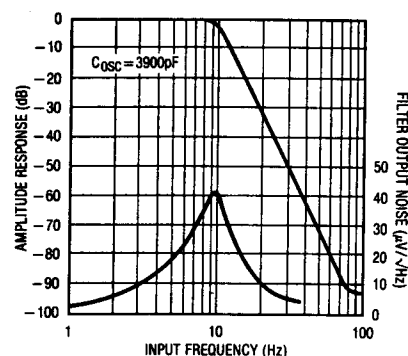
LTCMOS™ is a trademark of Linear Technology Corp.

TYPICAL APPLICATION

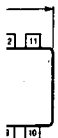


NOTE: TO ADJUST OSCILLATOR FREQUENCY, USE A 6800pF CAPACITOR IN SERIES WITH A 50K POT FROM PIN 5 TO GROUND.

Filter Amplitude Response and Noise



$$V_{OSZ} \left(\frac{R4}{R2} \right)$$

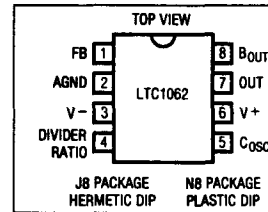


LTC1062

ABSOLUTE MAXIMUM RATINGS

Total Supply Voltage (V^+ to V^-) 18V
 Input Voltage at Any Pin $V^- - 0.3V \leq V_{IN} \leq V^+ + 0.3V$
 Operating Temperature Range
 LTC1062M $-55^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$
 LTC1062C $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$
 Storage Temperature Range -65°C to 150°C
 Lead Temperature Range (Soldering, 10 sec.) 300°C

PACKAGE/ORDER INFORMATION



ORDER PART NUMBER
LTC1062MJ8
LTC1062CJ8
LTC1062CN8

ELECTRICAL CHARACTERISTICS

Test Conditions: $V^+ = +5V$, $V^- = -5V$, $T_A = 25^\circ\text{C}$ unless otherwise specified, AC output measured at pin 7, Figure 1

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS	
Power Supply Current	C_{osc} (Pin 5 to V^-) = 100 pF		4.5	7 10	mA mA	
Input Frequency Range			0-20k		Hz	
Filter Gain at $f_{IN} = 0$	$f_{CLK} = 100\text{kHz}$, Pin 4 at V^+ $C = 0.01\mu\text{F}$, $R = 25.78\text{k}$		0	-0.02	-0.3	dB
$f_{IN} = 0.5f_C$ (Note 1)			-2	-3		dB
$f_{IN} = f_C$			-28	-30		dB
$f_{IN} = 2f_C$				-60		dB
$f_{IN} = 4f_C$						dB
Clock to Cutoff Frequency Ratio, f_{CLK}/f_C	Same as above		100 ± 1		%	
Filter Gain at $f_{IN} = 16\text{kHz}$	$f_{CLK} = 400\text{kHz}$, Pin 4 at V^+ $C = 0.01\mu\text{F}$, $R = 6.5\text{k}$	-45	-52		dB	
f_{CLK}/f_C Tempco	Same as above		10		ppm/°C	
Filter Output (Pin 7) DC Swing	Pin 7 buffered with an external op amp	± 3.5	± 3.8		V	
Clock Feedthrough			1		mVp-p	
Internal Buffer						
Bias Current			2	50	pA	
Bias Current			170	1000	pA	
Offset Voltage			2	20	mV	
Voltage Swing	$R_{load} = 20\text{k}\Omega$	± 3.5	± 3.8		V	
Short Circuit Current Source/Sink			40/3		mA	
Clock (Note 3)						
Internal Oscillator Frequency	C_{osc} (Pin 5 to V^-) = 100pF C_{osc} (Pin 5 to V^-) = 100pF	25 15	32	50 65	kHz kHz	
Max Clock Frequency			4		MHz	
Pin 5 Source or Sink Current			40	80	μA	

The ● denotes the specifications which apply over the full operating temperature range.

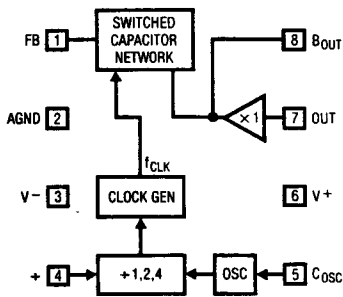
Note 1: f_C is the frequency where the gain is -3dB with respect to the input signal.

Note 2: The LTC1062M operates from $-55^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$, the LTC1062C operates from $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$.

Note 3: The external or driven clock frequency is divided by either 1, 2, or 4 depending upon the voltage at pin 4. When pin 4 = V^+ , ratio = 1; when pin 4 = GND, ratio = 2; when pin 4 = V^- , ratio = 4.

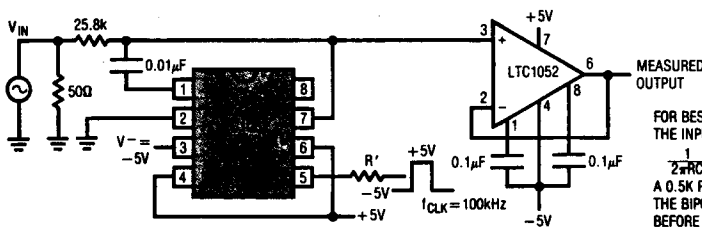
$$f_c = \frac{46,200}{33 + C} \quad 100:1$$

BLOCK DIAGRAM



BY CONNECTING PIN 4 TO V+, AGND, OR V-, THE OUTPUT FREQUENCY OF THE INTERNAL CLOCK GENERATOR IS THE OSCILLATOR FREQUENCY DIVIDED BY 1.2,4. THE (f_{CLK}/f_c) RATIO OF 100:1 IS WITH RESPECT TO THE INTERNAL CLOCK GENERATOR OUTPUT FREQUENCY. PIN 5 CAN BE DRIVEN WITH AN EXTERNAL CMOS LEVEL CLOCK. THE LTC1062 CAN ALSO BE SELF-CLOCKED BY CONNECTING AN EXTERNAL CAPACITOR (C_{OSC}) TO GROUND (OR TO V- IF C_{OSC} IS POLARIZED). UNDER THIS CONDITION AND WITH $\pm 5V$ SUPPLIES, THE INTERNAL OSCILLATOR FREQUENCY IS:
 $f_{OSC} = 140kHz [33pF / (33pF + C_{OSC})]$

For Adjusting Oscillator Frequency, Insert a 50K Pot in Series with C_{OSC} . Use Two Times Calculated C_{OSC} .

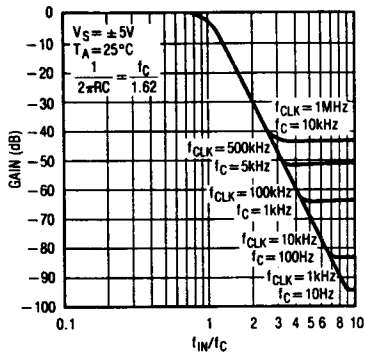


FOR BEST MAX. FLAT APPROXIMATION, THE INPUT R,C SHOULD BE SUCH AS:
 $\frac{1}{2\pi RC} = \frac{f_{CLK}}{100} \times \frac{1}{1.63}$
 A 0.5K RESISTOR, R', SHOULD BE USED IF THE BIPOLAR EXTERNAL CLOCK IS APPLIED BEFORE THE POWER SUPPLIES TURN ON.

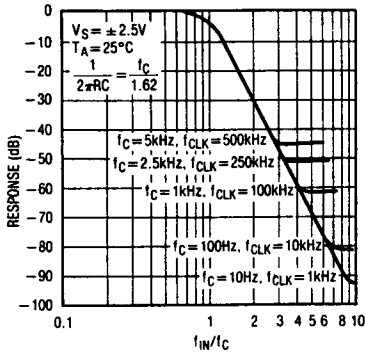
Figure 1. AC Test Circuit

TYPICAL PERFORMANCE CHARACTERISTICS

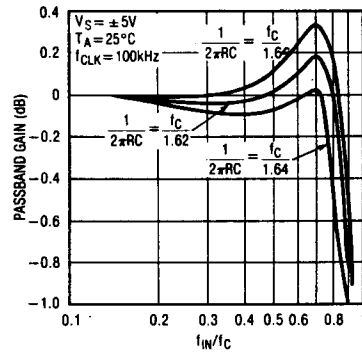
Amplitude Response Normalized to the Cutoff Frequency



Amplitude Response Normalized to the Cutoff Frequency



Passband Gain vs Input Frequency



ATION

ER
MBER

2MJ8
2CJ8
2CN8

UNITS

- mA
- mA
- Hz
- dB
- dB
- dB
- dB
- dB
- %
- dB

ppm/°C

/

p-p

pA

pA

mV

V

mA

KHz

KHz

MHz

μA

LTC1062C

or 1, 2, or 4

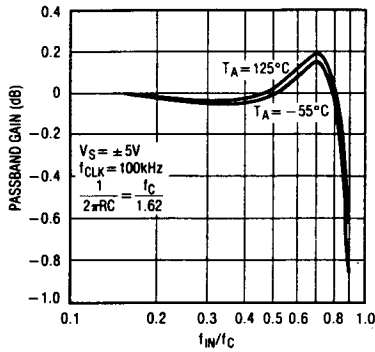
= 1; when

YEAR
TECHNOLOGY

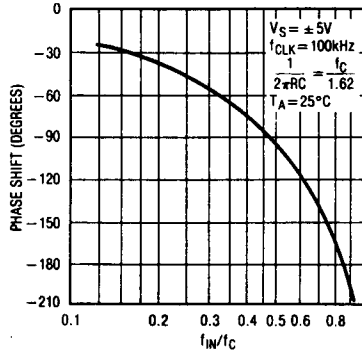


TYPICAL PERFORMANCE CHARACTERISTICS

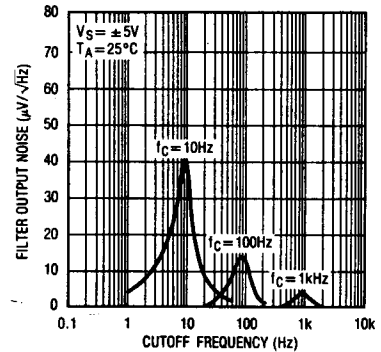
Passband Gain vs Input Frequency



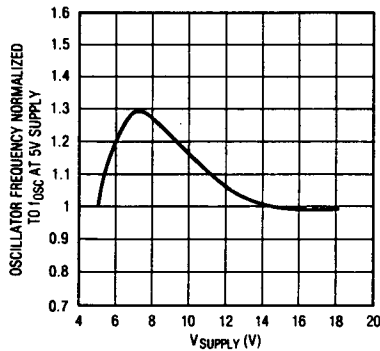
Passband Phase Shift vs Input Frequency



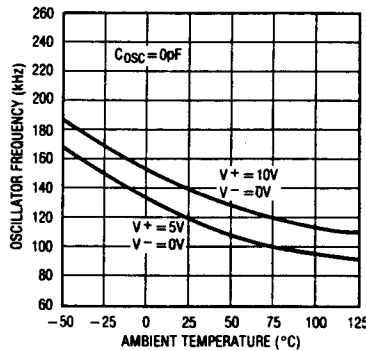
Filter Noise Spectral Density



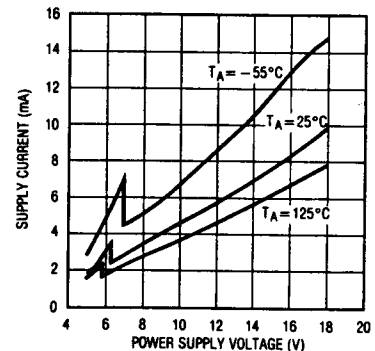
Normalized Oscillator Frequency, f_osc vs Supply Voltage



Oscillator Frequency, f_osc vs Ambient Temperature



Power Supply Current vs Power Supply Voltage



APPLICATIONS INFORMATION

Filter Input Voltage Range

Every node of the LTC1062 typically swings within 1V of either voltage supply, positive or negative. With the appropriate external (R,C) values, the amplitude response of all the internal or external nodes does not exceed a gain of 0 dB with the exception of pin 1. The amplitude response of the feedback node (pin 1) is shown in Figure 2. For an input frequency around $0.8 \times f_C$, the gain is 1.7 V/V and, with

$\pm 5V$ supplies, the peak-to-peak input voltage should not exceed 4.7V. If the input voltage goes beyond this value, clipping and distortion of the output waveform occur, but the filter will not get damaged nor will it oscillate. Also, the absolute maximum input voltage should not exceed the power supplies.

APPLICATIONS INFORMATION

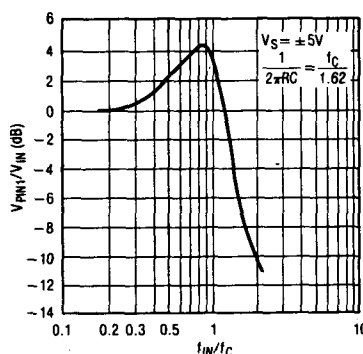


Figure 2. Amplitude Response of Pin 1

Internal Buffer

The internal buffer out (pin 8) and pin 1 are part of the signal AC path. Excessive capacitive loading will cause gain errors in the passband, especially around the cutoff frequency. The internal buffer gain at DC is typically 0.006dB. The internal buffer output can be used as a filter output, however it has a few millivolts of DC offset. The temperature coefficient of the internal buffer is typically $1\mu\text{V}/^\circ\text{C}$.

Filter Attenuation

The LTC1062 rolloff is typically 30dB/octave. When the clock, and the cutoff frequencies increase, the filter's maximum attenuation decreases. This is shown in the Typical Performance Characteristics. The decrease of the maximum attenuation, is due to the roll off at higher frequencies of the loop gains of the various internal feedback paths and not to the increase of the noise floor. For instance, for a 100kHz clock and 1kHz cutoff frequency, the maximum attenuation is about 64dB. A 4kHz, 1Vrms input signal will be predictably attenuated by 60dB at the output. A 6kHz, 1Vrms input signal will be attenuated by 64dB and not by 77dB as an ideal 5th order maximum flat filter would have dictated. The LTC1062 output at 6kHz will be about $630\mu\text{Vrms}$. The measured rms noise from DC to 17kHz was $100\mu\text{Vrms}$ which is 16dB below the filter output.

COSC, Pin 5

The C_{OSC} , pin 5, can be used with an external capacitor, C_{OSC} , connected from pin 5 to ground. If C_{OSC} is polarized it should be connected from pin 5 to the negative supply, pin 3. C_{OSC} lowers the internal oscillator frequency. If pin 5 is floating, an internal 33pF capacitor plus the external interpin capacitance set the oscillator frequency around 140kHz with $\pm 5\text{V}$ supply. An external C_{OSC} will bring the oscillator frequency down by the ratio $(33\text{pF})/(33\text{pF} + C_{OSC})$. The typical performance characteristics curves provide the necessary information to get the internal oscillator frequency for various power supply ranges. Pin 5 can also be driven with an external CMOS clock to override the internal oscillator. Although standard 7400 series CMOS gates do not guarantee CMOS levels with the current source and sink requirements of pin 5, they will, in reality, drive the C_{OSC} pin. CMOS gates conforming to standard B series output drive have the appropriate voltage levels and more than enough output current to simultaneously drive several LTC1062 C_{OSC} pins. The typical trip levels of the internal Schmitt trigger which input is pin 5, are given below.

V SUPPLY	V_{th+}	V_{th-}
$\pm 2.5\text{V}$	+0.9V	-1V
$\pm 5\text{V}$	+1.3V	-2.1V
$\pm 6\text{V}$	+1.7V	-2.5V
$\pm 7\text{V}$	+1.75V	-2.9V

APPLICATIONS INFORMATION

Divide By 1, 2, 4 (Pin 4)

By connecting pin 4 to V^+ , to mid supplies or to V^- , the clock frequency driving the internal switched capacitor network is the oscillator frequency divided by 1, 2, 4, respectively. Note that the f_{CLK}/f_C ratio of 100:1 is with respect to the internal clock generator output frequency. The internal divider is useful for applications where octave tuning is required. The ± 2 threshold is typically $\pm 1V$ from the mid supply voltage.

Transient Response

Figure 3 shows the LTC1062 response to a 1V input step.

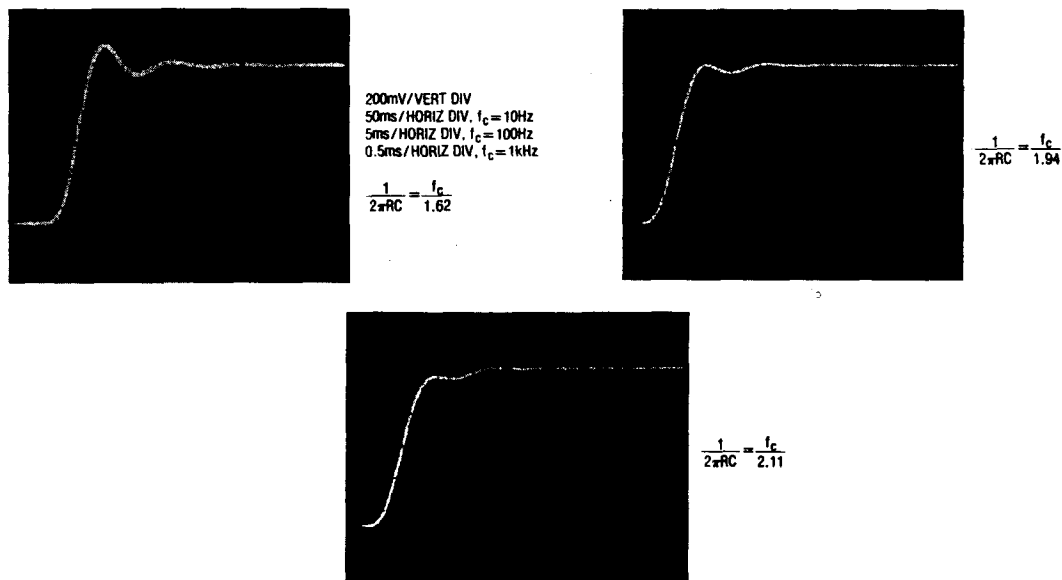


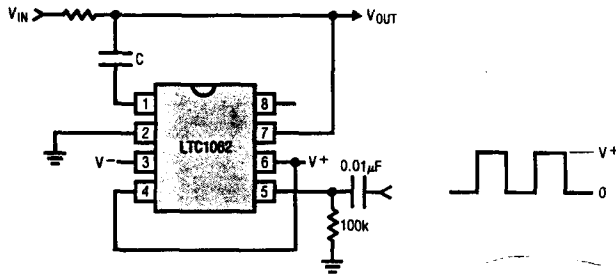
Figure 3. Step Response to a 1V Peak Input Step

Table 1

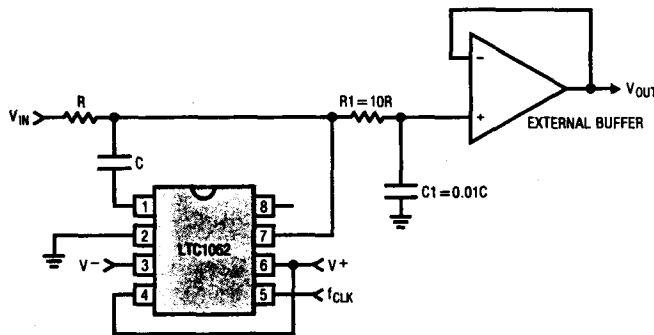
NOISE BW	rms NOISE $V_S = \pm 5V$
DC - $0.1 \times f_C$	2 μ V
DC - $0.25 \times f_C$	8 μ V
DC - $0.5 \times f_C$	20 μ V
DC - $1 \times f_C$	82 μ V
DC - $2 \times f_C$	100 μ V

TYPICAL APPLICATIONS

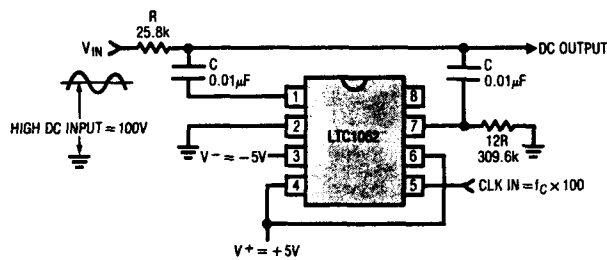
AC Coupling an External CMOS Clock Powered from a Single Positive Supply, V+



Adding an External (R1, C1) to Eliminate the Clock Feedthrough and to Improve the High Frequency Attenuation Floor



Filtering AC Signals from High DC Voltages

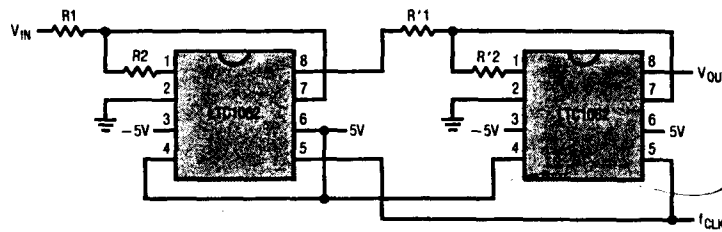


EX $f_{CLK} = 100\text{kHz}$, $f_C = 1\text{kHz}$. THE FILTER ACCURATELY PASSES THE HIGH DC INPUT AND ACTS AS 5th ORDER LP FILTER FOR THE AC SIGNALS RIDING ON THE DC. THE AMPLITUDE RESPONSE IN THE PASSBAND IS SHOWN BELOW.

LTC1062

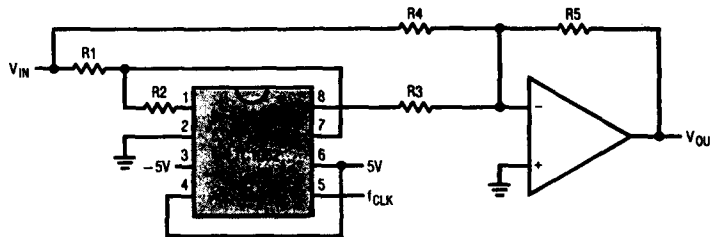
TYPICAL APPLICATIONS

Cascading Two LTC1062s to Form a Very Selective Clock Sweepable Bandpass Filter

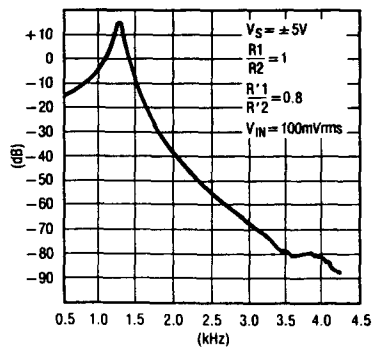


$R1 = 10k$, $R2 = 10k$
 $R'1 = 10k$, $R'2 = 12.5k$

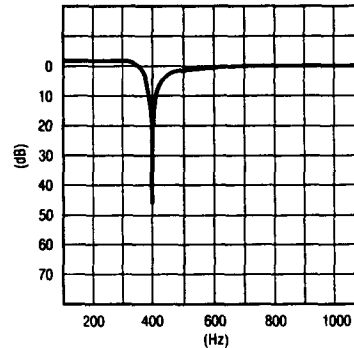
Clock Tunable Notch Filter
 For simplicity use $R3 = R4 = R5 = 10k$;
 $R1 = 1.234$, $f_{CLK} = \frac{79.3}{1}$
 $R2 = 1.234$, $f_{notch} = \frac{1}{1}$



Frequency Response of the Bandpass Filter

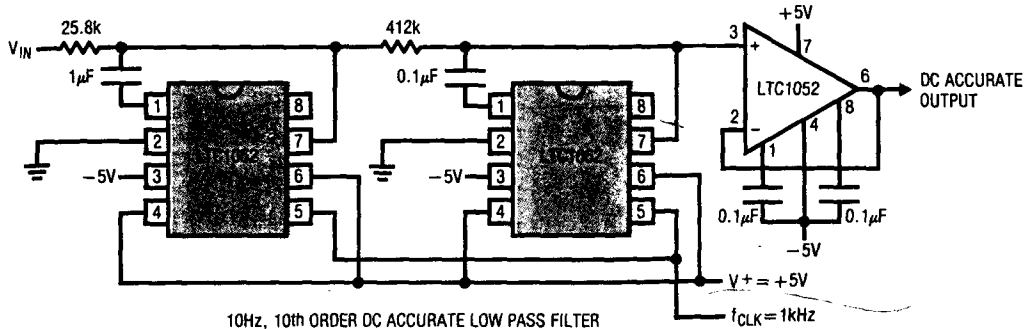


Notch Response



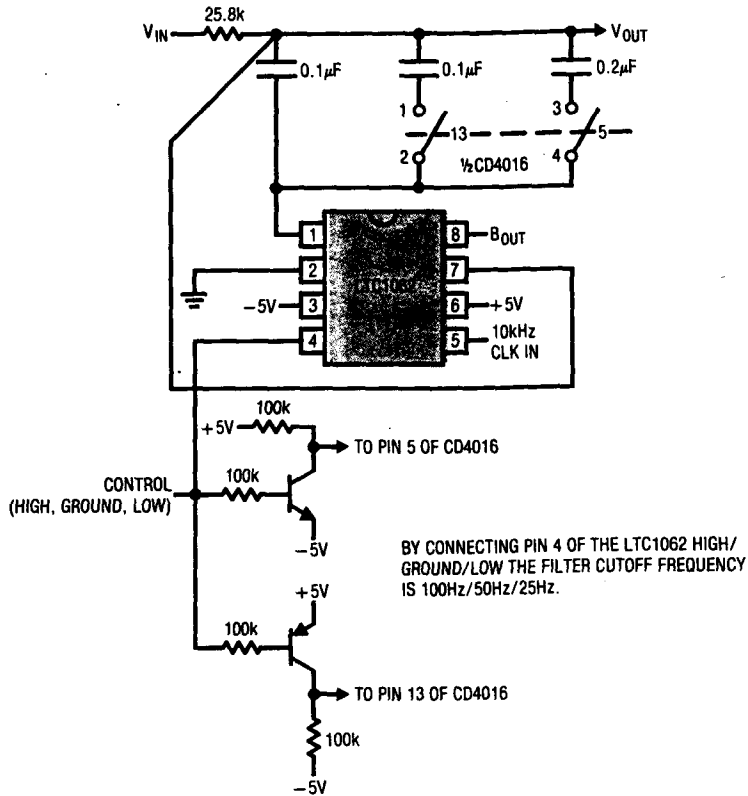
TYPICAL APPLICATIONS

Simple Cascading Technique



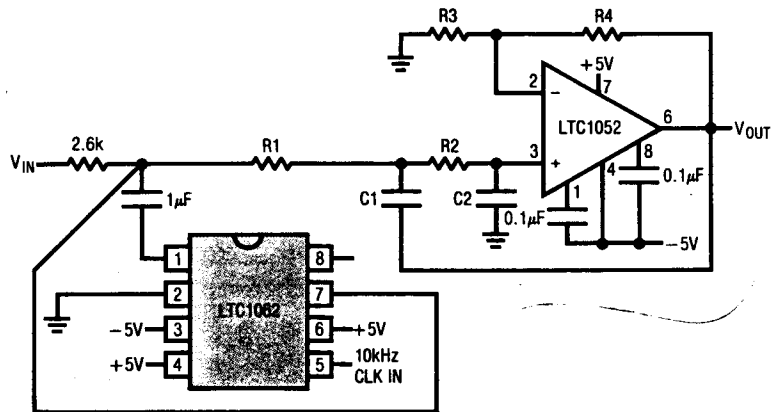
10Hz, 10th ORDER DC ACCURATE LOW PASS FILTER
 60dB/OCTAVE ROLLOFF
 0.5dB PASSBAND ERROR, 0dB DC GAIN
 MAXIMUM ATTENUATION 110dB ($f_{CLK}=10kHz$)
 100dB ($f_{CLK}=1kHz$)
 95dB ($f_{CLK}=1MHz$)

100Hz, 50Hz, 25Hz 5th Order DC Accurate LP Filter



TYPICAL APPLICATIONS

7th Order 100Hz Lowpass Filter with Continuous Output Filtering, Output Buffering and Gain Adjustment



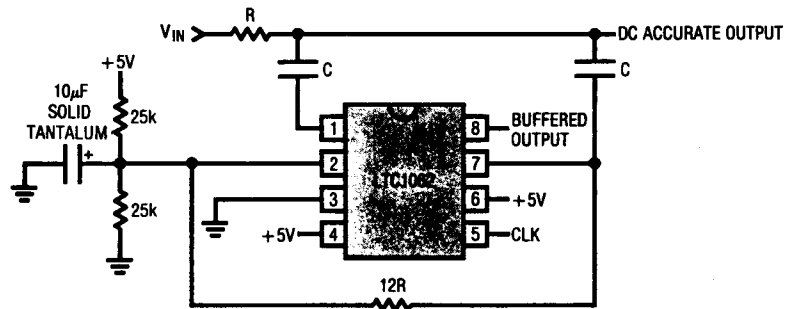
THE LTC1052 IS CONNECTED AS A 2nd ORDER SALLEN AND KEY LOWPASS FILTER WITH A CUTOFF FREQUENCY EQUAL TO THE LTC1062. THE ADDITIONAL FILTERING ELIMINATES ANY 10kHz CLOCK FEED THROUGH PLUS DECREASES THE WIDEBAND NOISE OF THE FILTER.

DC OUTPUT OFFSET (REFERRED TO A DC GAIN OF UNITY) = 5µV MAX.

WIDEBAND NOISE (REFERRED TO A DC GAIN OF UNITY) = 60µVrms

OUTPUT FILTER COMPONENT VALUES						
DC GAIN	R3	R4	R1	R2	C1	C2
1	∞	0	14.3k	53.6k	0.1µF	0.033µF
10	3.57k	32.4k	46k	274k	0.01µF	0.02µF

Single 5V Supply 5th Order LP Filter

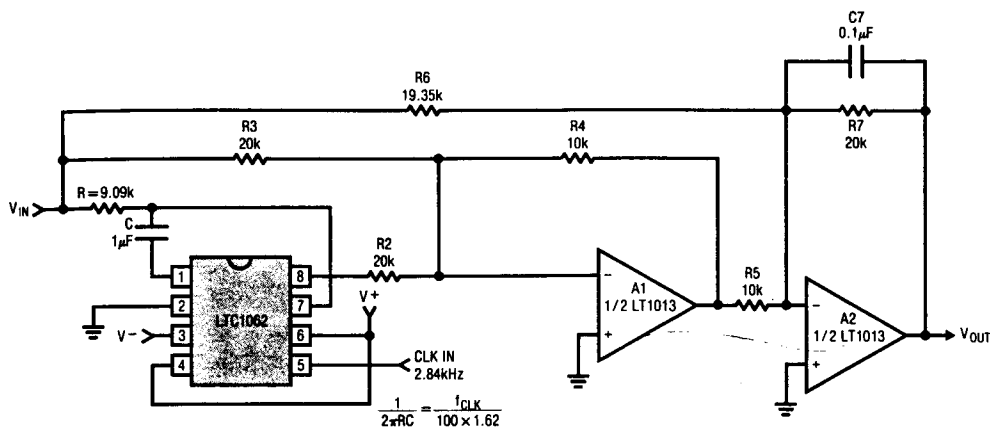


FOR A 10Hz FILTER R = 29.4k, C = 1µF, f_{CLK} = 1kHz

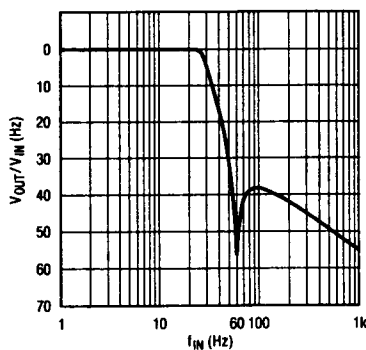
THE FILTER IS MAXIMALLY FLAT FOR $\frac{1}{2\pi RC} = \frac{f_c}{1.84}$

TYPICAL APPLICATIONS

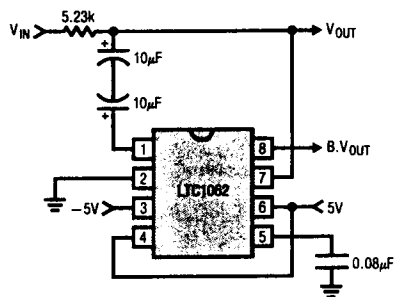
A Lowpass Filter with a 60Hz Notch



Frequency Response of the Above Lowpass Filter with the Notch $f_{NOTCH} = \frac{f_{CLK}}{47.3}$

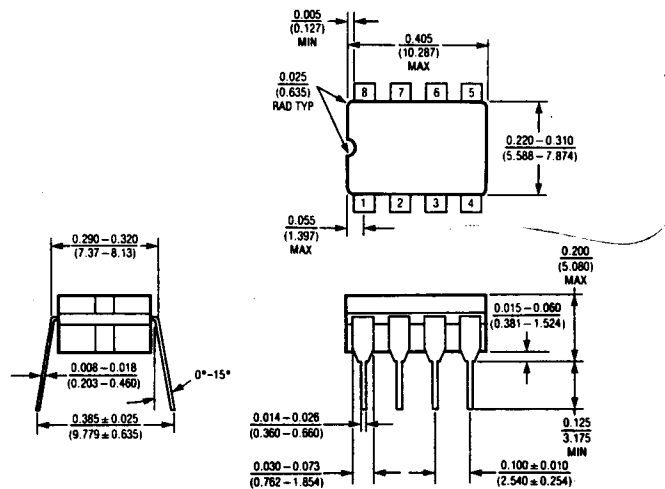


A Low Frequency, 5Hz Filter using Back-to-Back Solid Tantalum Capacitors



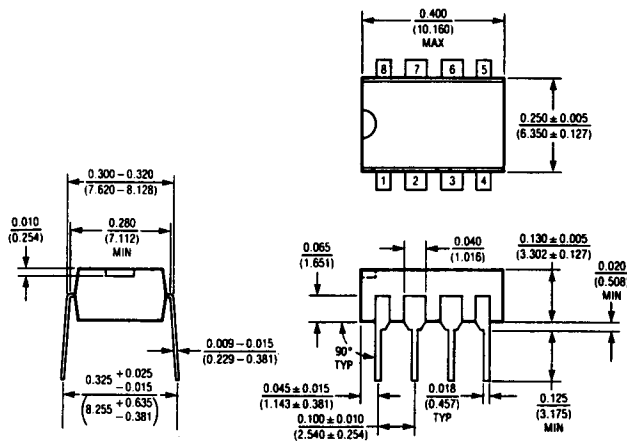
PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.

J8 Package
8 Lead Hermetic DIP



T_{jmax} 150°C	θ_{ja} 100°C/W
---------------------	--------------------------

N8 Package
8 Lead Plastic



T_{jmax} 100°C	θ_{ja} 130°C/W
---------------------	--------------------------